## ANALOG CIRCUITS LAB MANUAL



Department of Electronics \& Communication Engineering VEMU INSTITUTE OF TECHNOLOGY::P.KOTHAKOTA

NEAR PAKALA, CHITTOOR-517112
(Approved by AICTE, New Delhi \& Affiliated to JNTUA, Anantapuramu)

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Name: $\qquad$
H.T.No: $\qquad$

Year/Semester: $\qquad$

## Department of Electronics \& Communication Engineering

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# VEMU Institute of Technology <br> Dept. of Electronics and Communication Engineering 

## Vision of the institute

To be one of the premier institutes for professional education producing dynamic and vibrant force of technocrats with competent skills, innovative ideas and leadership qualities to serve the society with ethical and benevolent approach.

## Mission of the institute

Mission_1: To create a learning environment with state-of-the art infrastructure, well equipped laboratories, research facilities and qualified senior faculty to impart high quality technical education.
Mission_2: To facilitate the learners to inculcate competent research skills and innovative ideas by Industry-Institute Interaction.
Mission_3: To develop hard work, honesty, leadership qualities and sense of direction in learners by providing value based education.

## Vision of the department

To develop as a center of excellence in the Electronics and Communication Engineering field and produce graduates with Technical Skills, Competency, Quality, and Professional Ethics to meet the challenges of the Industry and evolving Society.

## Mission of the department

Mission_1: To enrich Technical Skills of students through Effective Teaching and Learning practices to exchange ideas and dissemination of knowledge.

Mission_2: To enable students to develop skill sets through adequate facilities, training on core and multidisciplinary technologies and Competency Enhancement Programs.

Mission_3: To provide training, instill creative thinking and research attitude to the students through Industry-Institute Interaction along with Professional Ethics and values.

## Programme Educational Objectives (PEOs)

PEO 1: To prepare the graduates to be able to plan, analyze and provide innovative ideas to investigate complex engineering problems of industry in the field of Electronics and Communication Engineering using contemporary design and simulation tools.

PEO-2: To provide students with solid fundamentals in core and multidisciplinary domain for successful implementation of engineering products and also to pursue higher studies.

PEO-3: To inculcate learners with professional and ethical attitude, effective communication skills, teamwork skills, and an ability to relate engineering issues to broader social context at work place

## Programme Outcomes(Pos)

| PO_1 | Engineering knowledge: Apply the knowledge of mathematics, science, engineering <br> fundamentals, and an engineering specialization to the solution of complex engineering problems. |
| :--- | :--- |
| PO_2 | Problem analysis: Identify, formulate, review research literature, and analyze complex <br> engineering problems reaching substantiated conclusions using first principles of mathematics, <br> natural sciences, and engineering sciences. |
| PO_3 | Design/development of solutions: Design solutions for complex engineering problems and <br> design system components or processes that meet the specified needs with appropriate <br> consideration for the public health and safety, and the cultural, societal, and environmental <br> considerations. |
| PO_4 | Conduct investigations of complex problems: Use research-based knowledge and research <br> methods including design of experiments, analysis and interpretation of data, and synthesis of the <br> information to provide valid conclusions. |
| PO_5 | Modern tool usage: Create, select, and apply appropriate techniques, resources, and modern <br> engineering and IT tools including prediction and modeling to complex engineering activities <br> with an understanding of the limitations. |
| PO_6 | The engineer and society: Apply reasoning informed by the contextual knowledge to assess <br> societal, health, safety, legal and cultural issues and the consequent responsibilities relevant to the <br> professional engineering practice. |
| PO_7 | Environment and sustainability: Understand the impact of the professional engineering <br> solutions in societal and environmental contexts, and demonstrate the knowledge of, and need <br> for sustainable development. |
| PO_8 | Ethics: Apply ethical principles and commit to professional ethics and responsibilities and norms <br> of the engineering practice. |
| PO_9 | Individual and team work: Function effectively as an individual, and as a member or leader in <br> diverse teams, and in multidisciplinary settings. |
| PO_10 | Communication: Communicate effectively on complex engineering activities with the <br> engineering community and with society at large, such as, being able to comprehend and write <br> effective reports and design documentation, make effective presentations, and give and receive <br> clear instructions. |
| PO_11 | Project management and finance: Demonstrate knowledge and understanding of the <br> engineering and management principles and apply these to one's own work, as a member and <br> leader in a team, to manage projects and in multidisciplinary environments. |
| $\mathbf{P O \_ 1 2 ~}$ | Life-long learning: Recognize the need for, and have the preparation and ability to engage in <br> independent and life-long learning in the broadest context of technological change. |

## Programme Specific Outcome(PSOs)

| PSO_1 | Higher Education : Qualify in competitive examination for pursuing higher education by <br> applying the fundamental concepts of Electronics and Communication Engineering domains such <br> as Analog \& Digital Electronics, Signal Processing, Communication \& Networking, Embeded <br> Systems, VLSI Design and Control systems etc., |
| :--- | :--- |
| PSO_2 | Employment: Get employed in allied industries through their proficiency in program specific <br> domain knowledge, Specalized software packages and Computer programming or became an <br> entrepreneur. |

## COURSE OUTCOMES $\left(\mathrm{CO}_{\mathbf{S})}\right.$

| CO No. | Description | Blooms <br> Level |
| :---: | :--- | :---: |
| C229.1 | Understand the Characteristics and Frequency response of various <br> amplifiers | 2 |
| C229.2 | Analyze and Design negative Feedback amplifiers, Oscillators and <br> Tuned amplifiers. | 4 |
| C229.3 | Determine the efficiency of various Power Amplifiers | 2 |

## LIST OF EXPERIMENTS:

1. Design and Analysis of Darlington pair.
2. Frequency response of $\mathrm{CE}-\mathrm{CC}$ multistage Amplifier
3. Design and Analysis of Cascode Amplifier.
4. Frequency Response of Differential Amplifier
5. Design and Analysis of Series - Series feedback amplifier and find the frequency response of it.
6. Design and Analysis of Shunt - Shunt feedback amplifier and find the frequency response of it.
7. Design and Analysis of Class A power amplifier
8. Design and Analysis of Class AB amplifier
9. Design and Analysis of RC phase shift oscillator
10. Design and Analysis of LC Oscillator
11. Frequency Response of Single Tuned amplifier
12. Design and Analysis of Bistable Multivibrator
13. Design and Analysis of Monostable Multivibrator
14. Design and Analysis of Astable Multivibrator
15. Design and Analysis of Darlington pair.
16. Design and Analysis of Cascode Amplifier.
17. Frequency Response of Differential Amplifier
18. Design and Analysis of Series - Series feedback amplifier and find the frequency response of it.
19. Design and Analysis of Shunt - Shunt feedback amplifier and find the frequency response of it.
20. Design and Analysis of Class A power amplifier
21. Design and Analysis of RC phase shift oscillator
22. Design and Analysis of LC Oscillator
23. Frequency Response of Single Tuned amplifier
24. Design and Analysis of Bistable Multivibrator
25. Design and Analysis of Monostable Multivibrator
26. Design and Analysis of Astable Multivibrator

## ADVANCED EXPERIMENTS:

1. Class B Push Pull Power Amplifier
2. Cascade Amplifier

## CONTENTS

| S.NO. | NAME OF THE EXPERIMENT | PAGE NO |
| :---: | :---: | :---: |
| 1 | Design and Analysis of Darlington pair. |  |
| 2 | Design and Analysis of Cascode Amplifier. |  |
| 3 | Frequency Response of Differential Amplifier |  |
| 4 | Design and Analysis of Series - Series feedback amplifier and find the frequency response of it. |  |
| 5 | Design and Analysis of Shunt - Shunt feedback amplifier and find the frequency response of it. |  |
| 6 | Design and Analysis of Class A power amplifier |  |
| 7 | Design and Analysis of RC phase shift oscillator |  |
| 8 | Design and Analysis of LC Oscillator |  |
| 9 | Frequency Response of Single Tuned amplifier |  |
| 10 | Design and Analysis of Bitable Multivibrator |  |
| 11 | Design and Analysis of Monostable Multivibrator |  |
| 12 | Design and Analysis of Astable Multivibrator |  |
| ADVANCED EXPERIMENTS |  |  |
| 1 | Class B Push Pull Power Amplifier |  |
| 2 | Cascade Amplifier |  |

## DOS \& DONTS IN LABORATORY

## DO's

1. Students should be punctual and regular to the laboratory.
2. Students should come to the lab in-time with proper dress code.
3. Students should maintain discipline all the time and obey the instructions.
4. Students should carry observation and record completed in all aspects.
5. Students should be at their concerned experiment table, unnecessary moment is restricted.
6. Students should follow the indent procedure to receive and deposit the components from lab technician.
7. While doing the experiments any failure/malfunction must be reported to the faculty.
8. Students should check the connections of circuit properly before switch ON the power supply.
9. Students should verify the reading with the help of the lab instructor after completion of experiment.
10. Students must endure that all switches are in the lab OFF position, all the connections are removed.
11. At the end of practical class the apparatus should be returned to the lab technician and take back the indent slip.
12. After completing your lab session SHUTDOWN the systems, TURNOFF the power switches and arrange the chairs properly.
13. Each experiment should be written in the record note book only after getting signature from the lab in charge in the observation notebook.

## DON'Ts

1. Don't eat and drink in the laboratory.
2. Don't touch electric wires.
3. Don't turn ON the circuit unless it is completed.
4. Avoid making loose connections.
5. Don't leave the lab without permission.
6. Don't bring mobiles into laboratory.
7. Do not open any irrelevant sites on computer.
8. Don't use a flash drive on computers.

SCHEME OF EVALUATION

| S.No | Program | Date | Marks Awarded |  |  |  | Total <br> 30(M) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\begin{gathered} \text { Record } \\ (10 \mathrm{M}) \end{gathered}$ | $\begin{gathered} \text { Obs. } \\ (\mathbf{1 0 M}) \end{gathered}$ | $\begin{aligned} & \text { Viva } \\ & (5 \mathrm{M}) \end{aligned}$ | Attd. (5M) |  |
| 1 | Design and Analysis of Darlington pair. |  |  |  |  |  |  |
| 2 | Design and Analysis of Cascode Amplifier. |  |  |  |  |  |  |
| 3 | Frequency Response of Differential Amplifier |  |  |  |  |  |  |
| 4 | Design and Analysis of Series <br> - Series feedback amplifier |  |  |  |  |  |  |
| 5 | Design and Analysis of Shunt <br> - Shunt feedback amplifier |  |  |  |  |  |  |
| 6 | Design and Analysis of Class <br> A power amplifier |  |  |  |  |  |  |
| 7 | Design and Analysis of RC phase shift oscillator |  |  |  |  |  |  |
| 8 | Design and Analysis of LC Oscillator |  |  |  |  |  |  |
| 9 | Frequency Response of Single Tuned amplifier |  |  |  |  |  |  |
| 10 | Design and Analysis of Bitable Multivibrator |  |  |  |  |  |  |
| 11 | Design and Analysis of Monostable Multivibrator |  |  |  |  |  |  |
| 12 | Design and Analysis of Astable Multivibrator |  |  |  |  |  |  |
| ADVANCED EXPERIMENTS |  |  |  |  |  |  |  |
| 1 | Class B Push Pull Power Amplifier |  |  |  |  |  |  |
| 2 | Cascade Amplifier |  |  |  |  |  |  |

## CIRCUIT DIAGRAM:

## Darlington Pair



## MODEL WAVE FORM:



## DARLINGTON PAIR

AIM: To design and construct a Darlington amplifier and to calculate the bandwidth and cut off Frequency.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | Transistor | BC -547 | 2 |
| 2 | Resistors | $100 \mathrm{~K} \Omega, 10 \mathrm{~K} \Omega$ | 2 |
| 3 | Capacitors | 1 uf | 2 |
| 4 | Function Generator | $0-3 \mathrm{MHz}$ | 1 |
| 5 | RPS | $0-30 \mathrm{~V}$ | 1 |
| 6 | CRO | $0-30 \mathrm{MHz}$ | 1 |
| 7 | Bread Board | - | 1 |
| 8 | Connecting Wires | - | As Per Required |

## OPERATION:

A Darlington transistor pair comprises of a couple of bipolar transistors that are coupled in order to deliver a very high-current gain from a low-base current. It is cascading of common collector-common collector. In this circuit, the emitter of the input transistor is connected to the base terminal of the output transistor. Therefore, the current that is amplified by the first transistor is again amplified by the second transistor. The current gain of single stage cc amplifier is high, so by using Darlington pair we can get high current gain.

## PROCEDURE:

1. Design the circuit for given specifications and connect the circuit.
2. Apply input signal to the circuit of $2 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}, 1 \mathrm{KHz}$ sine wave from the function generator with DC offset ON.
3. Tabulate amplitude of output signal with change in frequency with steps from 10 Hz to 1 MHz and determine gain using $\mathrm{A}_{\mathrm{v}}=\mathrm{V}_{0} / \mathrm{V}_{\mathrm{s}}$.
4. Calculate voltage gain in dB using $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{s}}\right)$.
5. Plot the frequency response on semi-log graph using the values of amplitude in dB Vs frequency in Hz .
6. Calculate the bandwidth using $B W=f_{2}-f_{1}$

## DESIGN OF DARLINGTON AMPLIFIER

Design parameters
$\mathrm{Vcc}=12 \mathrm{~V}, \mathrm{Ie}=2 \mathrm{~mA}$, hfe $(\beta)=100$, $\mathrm{Vbe}=0.7 \mathrm{~V}$,
$\mathrm{S}=10$,
Load resistance, $\mathrm{RL}=4.7 \mathrm{k} \Omega$
Design specifications
Vcc - Vce-Ve $=0$
$V_{C E}=50 \%$ of Vcc
Vce $=0.5 * 12=6 \mathrm{~V}$
VE=12-6
$\mathrm{VE}=6 \mathrm{~V}$
To find Reff $\mathrm{V}_{\mathrm{E}}=\mathrm{IE} \times \operatorname{Reff}^{\text {efeff }}=\mathrm{VE} / \mathrm{IE}$
$R_{\text {eff }}=6 / 2 \times 10-3$
Reff $=3 \mathrm{~K} \Omega$
To find $\operatorname{Re}_{\mathrm{E}} \mathrm{eff}=\mathrm{R}_{\mathrm{E}} \| \mathrm{R}_{\mathrm{L}}$
$3 \mathrm{~K} \Omega=\mathrm{RE} * 4.7 \mathrm{~K} \Omega / \mathrm{Re}+4.7 \mathrm{~K} \Omega$
$\operatorname{RE}=8.2 \mathrm{~K} \Omega$ use approx $10 \mathrm{~K} \Omega$
For determining the values of $R_{1} \& R_{2}\left(R_{B}=R_{1} \| R_{2}\right)$ following steps should be followed
Step 1 : Calculate Rв
Step 2 : Calculate Vtн
Let
$\mathrm{RB}_{\mathrm{B}}=\mathrm{R}_{1} \| \mathrm{R}_{2}$
$\mathrm{R}_{\mathrm{B}}=\mathrm{R}_{1} * \mathrm{R}_{2} / \mathrm{R}_{1}+\mathrm{R}_{2}$
VTH= Vcc*R2 / R1+R2-------------- (2)
Calculation of RB From Approx analysis $S=1+\left(R_{b} / R_{E}\right)$
$10=1+\mathrm{R}_{\mathrm{B}} / 8.2 \mathrm{~K} \Omega$
$9 * 8.2 \mathrm{~K} \Omega=\mathrm{R}_{\mathrm{B}} \mathrm{Rb}_{\mathrm{b}}=73 \mathrm{~K} \Omega$
Calculation of $\mathrm{V}_{\text {th }} \mathrm{V}_{\mathrm{th}}-\mathrm{V}_{\mathrm{be}}-\mathrm{Ve}_{\mathrm{E}}=0 \mathrm{~V}_{\mathrm{th}}=\mathrm{V}_{\mathrm{be}}+\mathrm{Ve}_{\mathrm{V}} \mathrm{V}_{\mathrm{th}}=0.7+6$
$\mathrm{V}_{\mathrm{TH}}=6.7 \mathrm{~V}$
From eqn (2)
$\mathrm{V}_{\mathrm{TH}} / \mathrm{Vcc}=\mathrm{R}_{2} / \mathrm{R}_{1}+\mathrm{R}_{2} 6.7 / 12=\mathrm{R}_{2} / \mathrm{R}_{1}+\mathrm{R}_{2}$
$0.558=\mathrm{R}_{2} / \mathrm{R}_{1}+\mathrm{R}_{2}-$
To find R1
From (1)
$\mathrm{R}_{\mathrm{B}}=\mathrm{R}_{1} * \mathrm{R}_{2} / \mathrm{R}_{1}+\mathrm{R}_{2} 7300=0.558 * \mathrm{R}_{1}$
$\mathrm{R}_{1}=130 \mathrm{k} \Omega$ use approx $150 \mathrm{k} \Omega$
To find $\mathrm{R}_{2}$
From (3)
$0.558=\mathrm{R}_{2} / \mathrm{R}_{1}+\mathrm{R}_{2} 0.558\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right)=\mathrm{R}_{2}$
$0.558\left(130 \times 103+\mathrm{R}_{2}\right)=\mathrm{R}_{2}$
$\mathrm{R}_{2}=162 \mathrm{k} \Omega$ use approx 150

OBSERVATIONS:

$$
V_{i}=2 \mathrm{~V}
$$

| S.NO | FREQUENCY(Hz) | OUTPUT <br> VOLTAGE (V) | GAIN $\left(\mathrm{V}_{0} / \mathrm{V}_{\mathrm{s}}\right)$ | GAIN IN dB <br> $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{s}}\right)$ |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |
|  |  |  |  |  |

## PRECUATIONS:

1. DC offset should be ON , to avoid clipping.

## CALCULATIONS :

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. How do you calculate the overall current gain of a Darlington pair?
2. What is the gain of Darlington transistor pair?
3. How do you make a Darlington pair?
4. What does a Darlington pair do?
5. Why is it called emitter follower?

## CIRCUIT DIAGRAM:

## CASCODE AMPLIFIER:



MODEL WAVE FORM:


Bandwidth $=\mathrm{f}_{\mathrm{H}}-\mathrm{f}_{\mathrm{L}}$

## $\operatorname{Exp} \mathrm{No}: 02$

## Date:

## CASCODE AMPLIFIER

AIM: To design and construct a cascade amplifier circuit and to draw its frequency response graph.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | Transistor | BC 107 or BC 547 | 2 |
| 2 | 2 | $33 \mathrm{~K} \Omega$ | 1 |
|  |  | Resistors | $22 \mathrm{~K} \Omega$ |
|  |  | $10 \mathrm{~K} \Omega$ | 1 |
|  |  | $1 \mathrm{~K} \Omega$ | 1 |
|  |  | $100 \Omega$ | 1 |
| 3 | Capacitors | $1 \mu \mathrm{~F}$ | 1 |
|  | RPS | $2.2 \mu \mathrm{~F}$ | 2 |
| 4 | Function Generator | $0-30 \mathrm{~V}$ | 1 |
| 5 | CRO | $0-3 \mathrm{MHz}$ | 1 |
| 6 | Bread board | 30 MHz | 1 |
| 7 | Connecting wires | - | 1 |
| 8 |  | - | 1 |

## OPERATION:

Cascode amplifier is a two stage circuit consisting of a transconductance amplifier followed by a buffer amplifier. The word "cascade" was originated from the phrase"cascade to cathode". This circuit have a lot of advantages over the single stage amplifier like, better input output isolation, better gain, improved bandwidth, higher input impedance, higher output impedance, better stability, higher slew rate etc. The reason behind the increase in bandwidth is the reduction of Miller effect. Cascode amplifier is generally constructed using FET ( field effect transistor) or BJT ( bipolar junction transistor). One stage will be usually wired in common source/common emitter mode and the other stage will be wired in common base/ common emitter mode

## PRECAUTIONS

Avoid loose connections give proper input voltage

## DESIGN :

Design parameters
$\mathrm{Vcc}=12 \mathrm{~V}, \mathrm{Ic}=2 \mathrm{~mA}$, hfe $(\beta)=100$, $\mathrm{Vbe}=0.7 \mathrm{~V}$,
$\mathrm{V}_{\text {CEI }}=\mathrm{V}_{\text {CE } 2}=35 \%$ of $\mathrm{Vcc}=4.2 \mathrm{~V}$
$V_{\text {Re }}=10 \%$ of $\mathrm{VCC}_{\mathrm{CC}}=1.2 \mathrm{~V} \mathrm{~V}_{\mathrm{RC}}=20 \%$ of $\mathrm{VCC}_{\mathrm{Cc}}=2.4 \mathrm{~V}$
To find Rc
$\mathrm{VRC}=\mathrm{Ic} * \mathrm{Rc}=2.4 \mathrm{~V}$
$\mathrm{Rc}=1.2 \mathrm{~K} \Omega$
To find Re
$V_{\mathrm{RE}}=\mathrm{IE} * \mathrm{Re}=1.2 \mathrm{~V}$ Re $=600 \Omega$
To find $\mathrm{R}_{1}, \mathrm{R}_{2}$ and $\mathrm{R}_{3}$
Vcc-Vri-Vbei-Vce2-Vre=0 Vri=Vcc-Vbei-Vce2-Vre Vri=12-0.6-4.2-12=6V
$\mathrm{IB}=\mathrm{Ic} / \mathrm{Hfe}=20 \mu \mathrm{~A}$
If 10 Ib assumed flowing through $\mathrm{R}_{1}$ we get
$\mathbf{R 1}=\mathrm{VR}_{\mathrm{R}} / \mathbf{1 0} \mathrm{IB}=\mathbf{3 0 K} \boldsymbol{K}$
Vcc-Vr1-Vr2-Vbe2-Vre=0 Vr2=Vcc- Vr1-Vbe2-Vre Vr2=12-6-0.6-1.2 $=4.2 \mathrm{~V}$
$\mathrm{IB}=\mathrm{Ic} / \mathrm{Hfe}=20 \mu \mathrm{~A}$
If 9Iв $_{\text {в }}$ assumed flowing through $\mathrm{R}_{2}$ we get
$\mathbf{R 2}=V_{R 2} / 9 \mathrm{Ib}=\mathbf{2 3 K} \Omega$
Vr3- $_{\text {- }}^{\text {be2 }}-\mathrm{V}_{\text {re }}=0$ Vr3 $=\mathrm{V}_{\mathrm{BE} 2}+\mathrm{V}_{\mathrm{RE}} \mathrm{V}_{\mathrm{R} 3}=0.6+1.2=1.8 \mathrm{~V}$
$\mathrm{IB}=\mathrm{Ic} / \mathrm{Hfe}=20 \mu \mathrm{~A}$
If 8 Ів assumed flowing through $\mathrm{R}_{3}$ we get
$R 3=V_{R 3} / 8 \mathrm{Ib}=11.2 \mathrm{~K} \Omega$
To find Ce (Bypass capacitor) $\mathrm{Xce}_{\mathrm{ce}}=\mathrm{Re} / 10$
$\mathrm{XCE}=600 \Omega / 10=60$
Xce $=1 / 2 \pi$ f CeLet $\mathrm{f}=1000$
$\mathrm{Ce}_{\mathrm{E}}=1 / 2 * \pi * 1000 * 60=2.2 \mu \mathrm{f}$

## PROCEDURE:

1. Connect the circuit diagram as shown in figure for cascode amplifier on breadboard.
2. Adjust input signal amplitude in the function generator and observe an amplified voltage at the output without distortion.
3. By keeping input signal voltage, say at 50 mV , vary the input signal frequency from 0 to 1 MHz in steps as shown in tabular column and note the corresponding output voltages.
4. Find the voltage gain, $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$
5. Plot AV VS frequency on a semi-log sheet.

## OBSERVATIONS:

| S.NO | FREQUENCY(Hz) | OUTPUT VOLTAGE (V) | GAIN(V0/Vi) | GAIN IN dB Av $=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |
|  |  |  |  |  |

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. Why cascode amplifier is used?
2. What are the features of cascode amplifier?
3. When amplifiers are cascaded the result is?
4. What is difference between Cascade and cascode?
5. Why cascode current mirrors are used?

## Circuit Diagram:

## DIFFERENTIAL AMPLIFIER:



## MODEL WAVE :



## DATE:

## DIFFERENTIAL AMPLIFIER

AIM: To design a differential amplifier and verify the frequency response.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | Transistor | $(\mathrm{BC107}$ or BC547 ) | 2 |
|  |  | 10 K | 4 |
| 2 | Resistors | 4.7 K | 1 |
|  | RPS | $0-12 \mathrm{~V}$ | 1 |
| 3 | Function Generator | $0-3 \mathrm{MHz}$ | 1 |
| 4 | CRO | 30 MHz | 1 |
| 5 | Bread board | - | 1 |
| 6 | Connecting wires | - | As Per Required |
| 7 |  |  |  |

## PROCEDURE:

1. Connections are given as per circuit diagram
2. Set $\mathrm{Vs}=50 \mathrm{mV}$, using signal generator
3. Keeping the input voltage constant vary the frequency from 50 Hz to 1 MHz in regular steps
4. Observe both input and output on the CRO (sine wave)
5. The differential gain is calculated at mid frequency range where the magnitude of the sine wave is maximum.
6. The differential gain is calculated by $\mathrm{Ad}=\mathrm{Vo} / \mathrm{Vi}$

## RECAUTIONS:

Avoid loose connections give proper input voltage

## Design Procedure:

$\mathrm{Vcc}=12 \mathrm{~V}$, Vee $=-12 \mathrm{~V}, \mathrm{Ic}_{1}=\mathrm{Ic}_{1}=2 \mathrm{~mA}, \mathrm{Ie}=4 \mathrm{~mA}, \mathrm{~h}_{\mathrm{fe}}(\beta)=300, \mathrm{Vbe}=0.7 \mathrm{~V}, \mathrm{~h}_{\mathrm{ie}}=4.7 \mathrm{k} \Omega$

## NOTE:

$\mathrm{Vcc}=12 \mathrm{~V}$
$V_{\mathrm{RE}}=10 \%$ of $\mathrm{Vcc}=0.1 * 12=1.2 \mathrm{~V}$
$\mathrm{V}_{\mathrm{RC}}=40 \%$ of $\mathrm{Vcc}=0.4 * 12=4.8 \mathrm{~V}$
$\mathrm{VcE}=50 \%$ of $\mathrm{Vcc}=0.5 * 12=6 \mathrm{~V} \mathrm{Ic} 1=\mathrm{Ic} 1=2 \mathrm{~mA}$

## To find Rc:

Apply KVL to collector loop
Vcc-IcRc-Vce-IeRe - Vee $=0$
Rc $=\{$ Vcc- Vce - Vre - Vee $\} /$ Ic
$=\{12-6-1.2-(-12)\} / 2 \times 10-3$
$\mathrm{Rc}=8.7 \mathrm{k} \Omega$ use approx $10 \mathrm{k} \Omega$

## To find Re:

Apply KVL to collector loop
Vcc-IcRc-Vce-IeRe - Vee $=0$
$\operatorname{Re}=\{$ Vcc- Vrc - Vce - Vee $\} /$ Ie
$=\{12-4.8-6-(-12)\} / 4 \times 10-3$
$\operatorname{Re}=3.3 \mathrm{k} \Omega$ use approx 4.7 k

## OBSERVATIONS:

| S.No | Frequency(Hz) | Output Voltage (V) | Gain=(V/Vi) | Gain in dB $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |
|  |  |  |  |  |
|  |  |  |  |  |
|  |  |  |  |  |

## MODEL GRAPH:



## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. What is CMRR?
2. What is Amplifier?
3. What Does Differential amplifier do?
4. what are the types of differential amplifier?
5. what are the advantages of differential amplifier?

## CIRCUIT DIAGRAM:

## Current Series Feedback Amplifier



## MODEL GRAPH:



## Date:

## CURRENT -SERIES (SERIES- SERIES) FEEDBACK AMPLIFIER

AIM: To design and test the current-series feedback amplifier and calculate the following parameters with feedback.

1. Mid band gain.
2. Bandwidth and cut-off frequencies.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | Transistor | BC 107 BP | 1 |
| 2 | Resistors | $22 \mathrm{~K} \Omega$ | 1 |
|  |  | $1 \mathrm{~K} \Omega$ | 1 |
|  |  | $4.7 \mathrm{~K} \Omega$ | 3 |
| 3 | Capacitors | $1 \mu \mathrm{~F}$ | 2 |
| 4 | RPS | $0-30 \mathrm{~V}$ | 1 |
| 5 | Function Generator | $0-3 \mathrm{MHz}$ | 1 |
| 6 | CRO | 30 MHz | 1 |
| 7 | Bread board | - | 1 |
| 8 | Connecting wires | - | As Per Required |

## PROCEDURE:

1. Connect the circuit as per the circuit diagram.
2. Keeping the input voltage constant, vary the frequency from 50 Hz to 3 MHz in regular steps and note down the corresponding output voltage.
3. Plot the graph: Gain (dB) Vs Frequency
4. Calculate the bandwidth from the graph.
5. Calculate the input and output impedance.
6. Remove Emitter Capacitance, and follow the same procedures (1 to 5).

## DESIGN PROCEDURE/ DESIGN CALCULATIONS:

$\mathrm{V}_{\mathrm{cc}}=12 \mathrm{v} ; \quad \mathrm{Ic}=1 \mathrm{~mA} ; \quad \mathrm{f}_{\mathrm{L}}=50 \mathrm{~Hz} ; \quad \mathrm{S}=2 ; \quad \mathrm{R}_{\mathrm{L}}=4.7 \mathrm{~K} \Omega$
$\mathrm{r}_{\mathrm{e}}=\frac{26 m \mathrm{~V}}{I c}=26 \Omega$
$\mathrm{V}_{\text {ce }}=\frac{V c c}{2}=6 \mathrm{~V}$
$\mathrm{V}_{\mathrm{E}}=V c c / 10=1.2 \mathrm{~V}$
$\mathrm{h}_{\text {ie }}=\mathrm{h}_{\mathrm{fe}} \mathrm{r}_{\mathrm{e}}=2.6 \mathrm{~K} \Omega$
Applying KVL output loop, we get

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{cc}}=\mathrm{I}_{\mathrm{E}} \mathrm{R}_{\mathrm{E}}+\mathrm{I}_{\mathrm{C}} \mathrm{R}_{\mathrm{C}}+\mathrm{V}_{\mathrm{ce}} ; \\
& 12=1 \times 10^{-3} \mathrm{x} \mathrm{R}_{\mathrm{C}}+6+1.2 \\
& \mathrm{R}_{\mathrm{C}}=4.8 \mathrm{~K} \Omega \\
& \mathrm{R}_{\mathrm{L}}=4.7 \mathrm{~K} \Omega
\end{aligned}
$$

Since $\mathrm{I}_{\mathrm{B}}$ is very small when compare with $\mathrm{I}_{\mathrm{C}}, \mathrm{I}_{\mathrm{C}} \approx \mathrm{I}_{\mathrm{E}}$

$$
\begin{aligned}
& \mathrm{R}_{\mathrm{E}}=\frac{V_{E}}{I_{E}}=1.2 \mathrm{~K} \Omega \\
& \mathrm{~S}=1+\frac{R_{B}}{R_{E}} \\
& 2=1+\left(\mathrm{R}_{\mathrm{B} / 1.2 \mathrm{~K})}\right. \\
& \mathrm{R}_{\mathrm{B}}=1.2 \mathrm{~K} \Omega \\
& \mathrm{~V}_{\mathrm{B}}=\frac{V_{C C} R_{2}}{R_{1}+R_{2}} \\
& 0.7=\frac{12 R 2}{R 1+R 2} \\
& \frac{R 2}{R 1+R 2}=0.05 \\
& \mathrm{R}_{\mathrm{B}}=\mathrm{R}_{1} \| \mathrm{R}_{2}
\end{aligned}
$$

By substituting above equation we will get
$\mathrm{R}_{1}=24 \mathrm{~K} \Omega$,
$\mathrm{R} 2=1.5 \mathrm{k} \Omega$
$\mathrm{X}_{\mathrm{Ci}}=\frac{h_{i e} \| R_{B}}{10}=0.00216 \mathrm{mho}$
$\mathrm{C}_{\mathrm{i}}=\frac{1}{2 \pi f X_{c i}}=1.47 \mu \mathrm{~F}$

Feedback factor, $\beta=-\mathrm{RE}_{\mathrm{E}}=1.2 \mathrm{k} \Omega$
$\mathrm{Gm}_{\mathrm{m}}=-\mathrm{hfe} /\left(\mathrm{hie}_{\mathrm{i}}+\mathrm{RE}\right)=-0.0263$

Desensitivity factor, $D=1+\beta G m=-1.63$

Transconductance with feedback, $\mathrm{Gmf}=\mathrm{Gm} / \mathrm{D}=0.01613$

## OBSERVATIONS:

$$
\mathrm{V}_{\mathrm{S}}=20 \mathrm{mv}
$$

| S.NO | Frequency(Hz) | Output Voltage (V) | Gain=( $\left.\mathrm{V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ | Gain in dB $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |
|  |  |  |  |  |

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1) What is feedback?
2) What is positive feedback?
3) Define amplification factor?
4) What is Q-point?
5) What is the effect of current series feedback amplifier on the input impedance of the amplifier?

## CIRCUIT DIAGRAM:

Voltage shunt Feedback Amplifier


## FREQUENCY RESPONSE

## Shunt Feedback Amplifier

Gain in dB


Exp No: 05

## Date:

## VOLTAGE SHUNT (SHUNT-SHUNT) FEEDBACK AMPLIFIER

AIM: To design and test the voltage shunt feedback amplifier and to calculate the following parameters with feedback.

1. Mid band gain.
2. Bandwidth and cut-off frequencies.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | Transistor | BC 107 | 1 |
| 2 | Resistors | $22 \mathrm{k} \Omega$ | 1 |
|  |  | $1.5 \mathrm{k} \Omega$ | 1 |
|  |  | $4.7 \mathrm{~K} \Omega$ | 2 |
|  |  | $1 \mathrm{~K} \Omega$ | 1 |
| 3 |  | $1 \mu \mathrm{~F}, 100 \mu \mathrm{~F}$ | 1 |
| 4 | Capacitors | $0-30 \mathrm{~V}$ | 1 |
| 5 | RPS | $0-3 \mathrm{MHz}$ | 1 |
| 6 | Function Generator | 30 MHz | 1 |
| 7 | CRO | - | 1 |
| 8 | Bread board | - | As Per Required |

## PROCEDURE:

1. Connect the circuit as per the circuit diagram.
2. Keeping the input voltage constant, vary the frequency from 50 Hz to 3 MHz in regular steps and note down the corresponding output voltage.
3. Plot the graph: Gain (dB) Vs Frequency
4. Calculate the bandwidth from the graph.
5. Calculate the input and output impedance.
6. Remove Emitter Capacitance, and follow the same procedures (1 to 5).

## DESIGN PROCEDURE/ DESIGN CALCULATIONS:

$\mathbf{V}_{\mathrm{cc}}=12 \mathrm{~V} \quad \mathrm{I}=1 \mathrm{~mA} ; \quad \mathrm{Av}_{\mathrm{C}}=30 ; \quad \mathrm{R}_{\mathrm{f}}=\mathbf{2 . 5} \mathrm{K} \Omega ; \quad \mathrm{s}=\mathbf{2} ;$
$\mathbf{r}_{\mathrm{e}}=\frac{26 m V}{I_{C}}=\mathbf{2 6 \Omega}$
$\beta=\frac{1}{R_{f}}=0.0004$
$\mathrm{h}_{\mathrm{fe}}=100$
$\mathrm{h}_{\mathrm{ie}}=\mathrm{h}_{\mathrm{fe}} \mathrm{r}_{\mathrm{e}} 2.6 \mathrm{k} \Omega$
$\mathrm{V}_{\mathrm{ce}}=\frac{V_{C C}}{2}=6 \mathbf{V}$
$\mathrm{V}_{\mathrm{E}}=\mathrm{Vcc} / 10=1.2 \mathrm{~V}$
Applying KVL to output loop, we get;
$\mathrm{V}_{\mathrm{cc}}=\mathrm{I}_{\mathrm{E}} \mathrm{R}_{\mathrm{E}}+\mathrm{I}_{\mathrm{C}} \mathrm{R}_{\mathrm{C}}+\mathrm{V}_{\mathrm{ce}}$;
$12=1 \times 10^{-3} \times \mathrm{R}_{\mathrm{C}}+6+1.2$
$\mathrm{R}_{\mathrm{C}}=4.8 \mathrm{~K} \Omega$
Since $I_{B}$ is very small when compare with $\mathrm{I}_{\mathrm{C}}, \mathrm{I}_{\mathrm{C}} \approx \mathrm{I}_{\mathrm{E}}$
$\mathrm{R}_{\mathrm{E}}=\frac{V_{E}}{I_{E}}$
$=1.2 \mathrm{k} \boldsymbol{\Omega}$
$\mathrm{S}=1+\frac{R_{B}}{R_{E}}$
$\mathbf{2 = 1}+\left(\mathrm{R}_{\mathrm{B} / 1.2 \mathrm{~K}}\right)$
$R_{B}=1.2 \mathrm{~K} \Omega$
$\mathrm{V}_{\mathrm{B}}=\frac{V_{C C} R_{2}}{R_{1}+R_{2}}$
$0.7=\frac{12 R 2}{R 1+R 2}$
$\frac{R 2}{R 1+R 2}=0.05$

$$
\mathbf{R}_{\mathbf{B}}=\mathbf{R}_{1} \| \mathbf{R}_{2}
$$

By substituting above equation we will get
$R_{1}=24 \mathrm{~K} \Omega, R 2=1.5 \mathrm{k} \Omega$
$\mathrm{Ro}=\mathrm{Rc} \| R_{f} R_{i}=\left(R_{B} \|\right.$ hie $) R_{f}=0.1895$
$R_{m}=-\left(h_{f e}\left(R_{B} \| R_{f}\right)\left(R_{C} \| R_{f}\right)\right) /\left(\left(R_{B} \| R_{f}\right)+h_{i e}\right)$

Desensitivity factor, $\mathrm{D}=1+\beta R_{m}$

$$
=1.0000758
$$

$R_{i f}=\frac{R i}{D}=2.19 \mathrm{k} \Omega$
$R_{o f}=\frac{R o}{D}=0.018 \Omega$
$\mathrm{R}_{\mathrm{mf}}=\frac{R m}{D}$
$\mathrm{X}_{\mathrm{ci}}=\frac{R_{i f}}{10}=219 \mathrm{mho}$
$\mathrm{C}_{\mathrm{i}}=\frac{1}{2 \pi f X c t}=0.1 \mu \mathrm{~F}$

## OBSERVATIONS:

$\mathrm{V}_{\mathrm{S}}=20 \mathrm{mv}$

| S.No | Frequency(hz) | Output Voltage (vo $)$ | Gain=(vo/vi) | Gain in db $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{v}_{0} / \mathrm{v}_{\mathrm{i}}\right)$ |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1) What is the difference between positive feedback and negative feedback?
2) What are the applications of feedback amplifiers?
3) Mention the properties of negative feedback?
4) Give an example of negative shunt feedback?
5) Define voltage shunt feedback?

## CIRCUIT DIAGRAM:

Class - A power amplifier


## MODELWAVE FORMS:



## Exp No: 06

## Date:

## CLASS - A POWER AMPLIFIER

AIM: To design a Class - A power amplifier using and compare practical efficiency with theoretical value.

## APPARATUS:

| S.NO | EQUIPMEN/COMPONENTS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | TRANSISTOR | $(\mathrm{SL} 100)$ | 1 |
| 2 | Resistors | $120 \mathrm{~K} \Omega$ | 1 |
|  |  | $560 \Omega$ | 1 |
|  |  | $470 \Omega$ | 1 |
| 3 |  | 22 uF | 2 |
| 4 | Capacitors | 50 mH | 1 |
| 5 | DIB | $0-50 \mathrm{~mA}$ | 1 |
| 6 | Ammeter | $0-30 \mathrm{~V}$ | 1 |
| 7 | RPS | 30 MHz | 1 |
| 8 | CRO | - | 1 |
| 9 | Bread Board | - | As Per Required |

## OPERATION:

The above-shown circuit is a directly coupled Class A amplifier. An amplifier where the load is coupled to the output of the transistor using a transformer is called a direct coupled amplifier.
Using transformer coupling technique, the efficiency of an amplifier can be enhanced to a great extent. The coupling transformer provides good impedance matching between the load and output, and it is the main reason behind the improved efficiency.

Generally, the current flows through the collector resistive load, this will cause the wastage of the DC power in it. As a result, this DC power dissipated in the load in a form of heat, and it does not contribute any output AC power.

Hence it is not advisable to pass the current through the output device (ex: loudspeaker) directly. For this reason, a special arrangement done by using a suitable transformer for coupling the load to the amplifier as given in the above circuit.

The circuit has the potential divider resistors R1 \& R2, biasing and emitter bypass resistor Re, used for circuit stabilization. The emitter bypass capacitor CE and emitter resistor Re are connected parallel to prevent AC voltage. The input capacitor Cin (Coupling Capacitor) used to couples AC input signal voltage to the base of the transistor and it blocks the DC from the previous stage.

## DESIGN PROCEDURE/ DESIGN CALCULATIONS

Let $\mathrm{V}_{\mathrm{cc}}=\mathbf{1 5 V}, \mathrm{I}_{\mathrm{c}}=\mathbf{1 6 m A}, \mathrm{C}=\mathbf{1 0 \mu F}$
$\mathrm{R}_{\mathrm{L}}=\frac{V c c}{2 I c}=468 \Omega \approx 470 \Omega$
$\mathrm{R}_{\mathrm{B}}=\frac{V c c-V b}{\text { Ibias }}=\frac{15-0.7}{0.16} \approx 56 \mathrm{~K} \Omega$
$I_{B}=\frac{I c}{\beta}=\frac{16 \mathrm{~mA}}{100}=0.16 \mathrm{~mA}$

## PROCEDURE:

1. Connect the circuit as shown in figure.
2. Adjust input signal amplitude in the function generator and observe an amplified voltage at the output without distortion.
3. By keeping input signal voltage, say at 150 mV , vary the input signal frequency 10 kHz as shown in tabular column and note the corresponding output voltage.
4. Measure and note down the zero signal dc current by disconnecting the function generator from the circuit.
5. Calculate the efficiency according to the expressions given.
6. Plot the graph between the o/p gain and frequency and calculate the bandwidth.

## CALCULATIONS

$\mathbf{P}_{\mathrm{dc}}=(\mathrm{Vcc})^{\mathbf{2}} / \mathbf{8 R} \mathrm{R}_{\mathrm{L}}$

$$
\mathbf{P}_{\mathrm{ac}}=\mathrm{Vcc} * \mathrm{Ic}
$$

$\% \boldsymbol{\eta}=\mathbf{P a c}_{\mathrm{a}} / \mathbf{P d c}{ }^{*} \mathbf{1 0 0}$

Efficiency $\mathbf{\%} \boldsymbol{\eta}=$
OBSERVATIONS:

$$
\mathrm{V}_{\mathrm{i}=150 \mathrm{mv}}
$$

| S.NO | FREQUENCY(Hz) | OUTPUT <br> VOLTAGE (V $\left.{ }_{0}\right)$ | GAIN <br> $\left(\mathrm{V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ | GAIN IN dB <br> $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |

## FREQUENCY RESPONSE



Bandwidth $=\mathrm{f}_{\mathrm{H}}-\mathrm{f}_{\mathrm{L}}$

## PRECAUTIONS:

1. No loose contacts at the junctions.
2. Check the connections before giving the power supply
3. Observations should be taken carefully.

## RESULT:

1. Frequency Response of CLASS-A Power amplifier is plotted.
2. Efficiency of CLASS A Power amplifier is found to be $\qquad$
3. Bandwidth $\mathrm{fH}-\mathrm{fL}=$ $\qquad$
4. Efficiency $(\% \eta)=$

CONCLUSION:

## VIVA QUESTIONS:

1. What is the main difference between general amplifier and power amplifier?
2. Why can't we get more current and voltage gains using general amplifier?
3. Is the power amplifier amplifies the power of input signal?
4. What are the classifications of Power amplifiers?
5. What does class A amplifier delivers?

## CIRCUIT DIAGRAM:

RC Phase Shift Oscillator:


## MODELWAVE FORMS:

## OUT PUT WAVEFORM:



OUTPUT WAVEFORM: $\theta=60^{\circ}$

## Date:

## RC PHASE SHIFT OSCILLATOR

AIM: To design a RC phase shift oscillator and verify practical frequency with theoretical Frequency.

## APPARATUS:

| S.NO | EQUIPMEN/COMPONENTS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | TRANSISTOR | BC547 | 1 |
| 2 | Resistor | $1 \mathrm{k} \Omega$ | 1 |
|  |  | $220 \Omega$ | 3 |
|  |  | $0.1 \mu \mathrm{f}$ | 3 |
| 3 | Capacitors | - | 1 |
| 4 | DRB | $0-30 \mathrm{~V}$ | 1 |
| 5 | RPS | 30 MHz | 1 |
| 6 | CRO | - | 1 |
| 7 | Bread Board | - | As Per Required |
| 8 | Connecting Wires |  |  |

## OPERATION:

When the circuit is switched on, current through R3 starts increasing because of biasing. This charging current induces voltage across R2 through C3.
The voltage across R2 leads the voltage across R3 by $60^{\circ}$.since three R-C sections are provided, therefore, the phase shift circuit produces a total phase shift of $60 \times 3=180^{0}$.
A further phase shift of $180^{\circ}$ is produced due to the transistor properties. So a total shift of 360 degrees is produced.
Therefore a fraction of the output fed to the input is in phase with it.
The RC phase shift oscillator can be made variable by making the resistors or capacitors variable. The common approach is to leave the resistors untouched the three capacitors are replaced by a triple gang variable capacitor.

## PROCEDURE:

1. Connections are made as per the circuit diagram.
2. Switch ON the power supply and set the biasing voltage $\mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V}$.
3. Adjust the $10 \mathrm{~K} \Omega$ pot to get a stable sinusoidal output and observe the sine wave form on oscilloscope.
4. Measure the frequency of oscillations of the output from the oscilloscope, then compare with theoretical value.
5. With respect to the output Vo, the waveforms at points $\mathrm{TP}_{1}, \mathrm{TP}_{2}$ and $\mathrm{TP}_{3}$, are observed on oscilloscope. We can see the phase shift at each point being shifted by an angle $600,1200,1800$.
6. Draw the waveform on graph sheet.

## DESIGN:

Amplifier design
Let $\mathrm{Vcc}=12 \mathrm{~V}, \mathrm{Ic}=4 \mathrm{~mA}, \mathrm{hf}_{\mathrm{fe}}=100$
Let $\mathrm{V}_{\mathrm{E}}=2 \mathrm{~V}, \mathrm{Vce}=6 \mathrm{~V}$
Therefore $\mathrm{R}_{\mathrm{E}}=\mathrm{V}_{\mathrm{E}} / \mathrm{I}_{\mathrm{E}}$

$$
=V_{E} / I_{c}
$$

$$
=2 / 4 \mathrm{~mA}
$$

$$
=0.5 \mathrm{~K} \Omega=500 \Omega
$$

So use $\mathrm{R}_{\mathrm{E}}=470 \Omega$

Apply KVL to the CE Loop
$\mathrm{V}_{\mathrm{cc}}-\mathrm{I}_{\mathrm{c}} \mathrm{Rc}_{\mathrm{c}}-\mathrm{V}_{\mathrm{CE}}-\mathrm{V}_{\mathrm{E}}=0$
$12-4 \mathrm{R}_{\mathrm{c}}-6-2=0$
Therefore $\mathrm{R}_{\mathrm{c}}=1 \mathrm{k} \Omega$

## Calculation of $\mathbf{R}_{1}$ AND $\mathbf{R}_{2}$

From biasing circuit
$\mathrm{V}_{\mathrm{B}}=\mathrm{V}_{\mathrm{cc}} \frac{R 2}{R 1+R 2}$
We know that
$\mathrm{V}_{\mathrm{B}}=\mathrm{V}_{\mathrm{Be}}+\mathrm{VE}=2+0.7$

$$
=2.7 \mathrm{~V}
$$

Therefore $\quad \frac{V B}{V c c}=\frac{R 2}{R 1+R 2}$

$$
\frac{2.7}{12}=\frac{R 2}{R 1+R 2}
$$

$\frac{R 2}{R 1+R 2}=0.225$
$0.225 \mathrm{R}_{1}+0.225 \mathrm{R}_{2}=\mathrm{R}_{2}$
$\mathrm{R}_{1}=3.44 \mathrm{R}_{2}$
If $\mathrm{R}_{2}=6.8 \mathrm{k} \Omega$, then $\mathrm{R}_{1}=23.3 \mathrm{k} \Omega$

Use $\mathrm{R}_{1}=22 \mathrm{k} \Omega$
Use $C_{E}=50 \mu \mathrm{~F}$ or $47 \mu \mathrm{~F}$
Also $\mathrm{C}_{\mathrm{c}}=0.1 \mu \mathrm{~F}$

## Design of shifting network

The frequency of oscillations is determined by phase shifting network. The oscillating frequency for the above circuit is given by

$$
f_{0}=\frac{1}{2 \pi R C \sqrt{6+4 K}}
$$

Where $k=\frac{R_{c}}{R}$ which is usually $<\mathrm{I}$
Let $f_{0}=2 \mathrm{KHz}$ (Audio frequency range 20 Hz to 20 KHz ) and $\mathrm{R}=2.2 \mathrm{~K} \Omega$
Therefore $K=\frac{R_{c}}{R}=\frac{1 K}{2.2 K}=0.454$
Therefore

$$
f_{0}=\frac{1}{2 \pi R C \sqrt{6+4(0.454)}}
$$

$\mathrm{C}=0.0121 \mu \mathrm{~F}$; use $\mathrm{C}=0.01 \mu \mathrm{~F}$

$$
h_{f e(\text { min })}=23+29 \times \frac{R}{R_{c}}+4 \times \frac{R_{c}}{R}
$$

Where $\mathrm{R}_{\mathrm{c}}=1 \mathrm{~K} \Omega$ and $\mathrm{R}=2.2 \mathrm{~K} \Omega$ (Phase shifting network)

$$
\begin{gathered}
h_{f e(\text { min })}=23+29 \times \frac{2.2 K}{1 K}+4 \times \frac{1 K}{2.2 K} \\
h_{f e(\text { min })}=89
\end{gathered}
$$

## THEORETICAL CALCULATIONS:

$\mathrm{C}=0.1 \mu \mathrm{~F}, \mathrm{R}=220 \Omega$

$$
f_{0}=\frac{1}{2 \pi R C \sqrt{6}}
$$




## OBSERVATIONS:

| S.NO | RESISTANCE | CAPACITANCE | AMPLITUDE(V) | TIME <br> PERIOD(ms) | TF | PF |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |

## RESULT:

Theoretical frequency of oscillations $=$ $\qquad$ KHz

Practical frequency of oscillations = $\qquad$ KH

## CONCLUSION:

## VIVA QUESTIONS:

1. How does RC phase shift oscillator work?
2. What is RC phase shift?
3. What is the basic principle of oscillator?
4. What are the advantages of RC phase shift oscillator?
5. What is the gain of RC phase shift oscillator?

CIRCUIT DIAGRAM:


Fig: Hartley Oscillator
MODEL WAVE FORM:


## Date:

## HARTLEY OSCILLATOR

AIM: To determine the frequency of the Hartley oscillator and verify it to the theoretical value

## APPARATUS:

| S.NO | EQUIPMEN/COMPONENTS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | TRANSISTOR | BC547 | 1 |
|  |  | $15 \mathrm{~K} \Omega$ | 1 |
|  |  | $2.7 \mathrm{~K} \Omega$ | 1 |
| 2 | Resistors | $1 \mathrm{~K} \Omega$ | 1 |
|  |  | $330 \Omega$ | 1 |
|  |  | $270 \Omega$ | 1 |
|  |  | $0.47 \mu \mathrm{~F}$ | 2 |
| 3 | Capacitors | 500PF | 1 |
| 4 | DIB | 100 mH | 2 |
| 5 | DCB | 500pf | 1 |
| 6 | RPS | 0-30V | 1 |
| 7 | CRO | $0-30 \mathrm{MHz}$ | 1 |
| 8 | Bread Board | - | 1 |
| 9 | Connecting Wires | - | As Per Required |

## OPERATION:

When the power supply is switched ON the transistor starts conducting and the collector current increases. As a result the capacitor C 1 starts charging and when the capacitor C 1 is fully charged it starts discharging through coil L1. This charging and discharging creates a series of damped oscillations in the tank circuit and it is the key. The oscillations produced in the tank circuit is coupled (fed back) to the base of Q1 and it appears in the amplified form across the collector and emitter of the transistor.

The output voltage of the transistor (voltage across collector and emitter) will be in phase with the voltage across inductor L1. Since the junction of two inductors is grounded, the voltage across L2 will be $180^{\circ}$ out of phase to that of the voltage across L1. The voltage across L2 is actually fed back to the base of Q1. From this we can see that, the feed back voltage is $180^{\circ}$ out of phase with the transistor and also the transistor itself will create another $180^{\circ}$ phase difference. So the total phase difference between input and output is $360^{\circ}$ and it is very important condition for creating sustained oscillations.
The frequency " $F$ " of a Hartley oscillator can be expressed using the equation;

$$
f=\frac{1}{2 \pi \sqrt{L C}}
$$

C is the capacitance of the capacitor C 1 in the tank circuit.
$\mathrm{L}=\mathrm{L} 1+\mathrm{L} 2$, the effective series inductance of the inductors L 1 and L 2 in the tank circuit.
Here the coils L1 and L2 are assumed to be winded on different cores. If they are winded on a single core then $\mathrm{L}=\mathrm{L} 1+\mathrm{L} 2+2 \mathrm{M}$ where M is the mutual inductance between the two coils.

## DESIGN PROCEDURE/ DESIGN CALCULATIONS:

## Transistor: BC 107

Let $\mathrm{Vcc}=12 \mathrm{~V} ;$ Ic $=4.5 \mathrm{~mA} ; \mathrm{VE}=1.2 \mathrm{~V} ; \mathrm{VcE}=6 \mathrm{~V}$;
$h$ Fe $=100$.
Given $\mathrm{Ve}_{\mathrm{E}}=1.2 \mathrm{~V}$. Therefore $\mathrm{Re}=\mathrm{VE}_{\mathrm{E}} / \mathrm{IE} \approx \mathrm{Ve}_{\mathrm{E}} / \mathrm{Ic}=266.67 \Omega ; \mathbf{R e}_{\mathrm{E}}=\mathbf{2 7 0 \Omega}$
Writing KVL for the Collector loop we get, $\mathrm{V}_{\mathrm{Cc}}=\mathrm{IcRc}+\mathrm{V}_{\mathrm{CE}}+\mathrm{Ve}_{\mathrm{E}}$
$\mathrm{R}_{\mathrm{C}}=\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{CE}}-\mathrm{V}_{\mathrm{E}}\right) / \mathrm{IC}=(12-6-1.2) \mathrm{V} / 4 \mathrm{~mA}=1.06 \mathrm{~K} \Omega ; \mathbf{R c}=\mathbf{1} \mathbf{K} \Omega$
$h_{F E} \operatorname{Re}=10 R_{2}$
Assume $\mathbf{R}_{\mathbf{2}}=\mathbf{2 . 7} \mathbf{K} \boldsymbol{\Omega}$,
$\mathrm{V}_{\mathrm{B}}=\left(\mathrm{V}_{\mathrm{cc}} \mathrm{x} \mathrm{R}_{2}\right) /\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right)$
Hence $\mathrm{R} 1=14.14 \mathrm{~K} \Omega ; \mathbf{R 1}=\mathbf{1 5} \mathrm{K} \Omega$

Use Ccı $=\mathbf{0 . 4 7 \mu F}$
Use Cc2 $=\mathbf{0 . 4 7} \boldsymbol{\mu} \mathbf{F}$
Use $\mathbf{C e}=\mathbf{4 7 \mu F}$

## Hartley Oscillator

Oscillator Frequency $f=c$. Leq. $=c .\left(L_{1}+L_{2}\right)$
Assume $f=500 \mathrm{KHz}$. With $L_{1}=L_{2}=100 \mu \mathrm{H}$,
we get
Leq. $=L_{1}+L_{2}=200 \mu \mathrm{H}$
Leq. $C=1 /(2 \pi f)^{2}=(\pi)^{-2} \times 10^{-12}$
This gives $\mathrm{C}=\left\{1 /(\pi)^{2} \times 200 \mu \mathrm{H}\right\} \mathrm{pF}=500 \mathrm{pF}$
Use $\mathrm{C}=470 \mathrm{pF}$
For this capacitance value $\boldsymbol{f}=\mathbf{5 1 8 . 6} \mathbf{K H z}$

## THEORETICAL CALCULATIONS:

$\mathrm{C}=500 \mathrm{pF}, \mathrm{LT}=100 \mathrm{mH}+100 \mathrm{mH}\left(\mathrm{L}_{1}+\mathrm{L}_{2}\right)$

$$
f=\frac{1}{2 \pi \sqrt{L C}}
$$

## PROCEDURE:

1. Connect the circuit diagram as shown in the figure
2. Set $\mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V}$
3. Keep the capacitance of the decade capacitance box, and measure the generated output signal amplitude and frequency from CRO.
4. Vary the Inductance in steps and note down frequency and amplitude at each.
5. Plot the graph from CRO and verify the practical frequency with theoretical frequency.

## TABULAR COLUMN:

| $\mathrm{C}(\mu \mathrm{f})$ | $\mathrm{L}_{1}$ <br> $(\mathrm{mH})$ | $\mathrm{L}_{2}$ <br> $(\mathrm{mH})$ | Theoretical <br> $\mathrm{F}(\mathrm{hz})$ | Practical <br> Time period | Practical <br> F(hz) | Amplitude (v) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. What is the use of Hartley oscillator?
2. What are the advantages of Hartley oscillator?
3. Recommended frequency range of Harley oscillator is?
4. Which component of Hartley oscillator is used in the feedback system?
5. Which network is used to give feedback to transistor of Hartley oscillator?

CIRCUIT DIAGRAM: SINGLE TUNED VOLTAGE AMPLIFIER


MODEL GRAPH:


## Date:

## SINGLE TUNED VOLTAGE AMPLIFIER

AIM: To design single tuned voltage amplifier and verify practical frequency with theoretical frequency. APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :--- | :---: | :---: |
| 1 | Transistor | BC 107 BP | 1 |
| 2 |  | Resistors | $100 \mathrm{~K} \Omega$ |
|  |  | $56 \mathrm{~K} \Omega$ | 1 |
|  |  | $33 \mathrm{~K} \Omega$ | 1 |
|  |  | $560 \Omega$ | 1 |
| 3 | Capacitors | $1 \mu \mathrm{~F}$ | 1 |
|  |  | $0.1 \mu \mathrm{~F}$ | 2 |
|  |  | 100 mH | 1 |
| 6 | RPS | $0-30 \mathrm{MHz}$ | 1 |
| 7 | CRO | $0-30 \mathrm{~V}$ | 1 |
| 8 | Bread board | 30 MHz | 1 |
| 9 | Connecting wires | - | 1 |

## OPERATION:

The circuit operation of single tuned amplifiers begins with the application of the high-frequency signal that is to be amplified at the base-emitter terminal of the transistor, shown in the figure above. By varying the capacitor employed in the tuned circuit, the resonant frequency of the circuit can be made equivalent to the frequency of the applied input signal. Here, the high impedance is offered to the signal frequency by the tuned circuit. Thus, a large output is achieved. For an input signal with multiple frequencies, only the frequency that corresponds to resonant frequency will get amplified. While all other frequencies are rejected the LC circuit. Hence, only the desired frequency signal gets selected and thus amplified by the circuit.

## CALCULATIONS:

$\mathrm{L}=100 \mathrm{mH}, \mathrm{C}=0.1 \mu \mathrm{~F} \quad$ Theoretical frequency $\mathrm{f}=1 / 2 \pi \sqrt{ } \mathrm{C} \mathrm{L}$

## PROCEDURE:

1. Connections are made as per the circuit diagram.
2. A signal of 1 KHz frequency and 50 mV peak-to-peak of sine wave is applied at the Input of amplifier.
3. By keeping the input voltage constant, vary the frequency from 1 KHz to 2 KHz in regular steps and note down the corresponding output voltage.
4. Calculate practically the frequency of oscillations by using the expression.
$\mathrm{f}=1 / \mathrm{T}_{\mathrm{d}}\left(\mathrm{T}_{\mathrm{d}}=\right.$ Time period ) and compare it with the theoretical frequency $\mathrm{f}=1 / 2 \pi \sqrt{ } \mathrm{C} L$
5. Voltage gain in dB is calculated by using the expression $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V} 0 / \mathrm{V}_{\mathrm{i}}\right)$
6. Plot graph for gain $\left(\mathrm{A}_{\mathrm{v}}\right)$ in dB vs frequency in Hz on a semi log graph.
7. The Bandwidth of the amplifier is calculated from the graph using the Expression,

Bandwidth $\mathrm{BW}=\mathrm{f}_{2}-\mathrm{f}_{1}$
Where $f_{1}$ is lower 3 dB frequency, f 2 is upper 3 dB frequency

## OBSERVATIONS:

$$
\mathrm{V}_{\mathrm{i}}=20 \mathrm{mV}
$$

| S.NO | FREQUENCY(Hz) | OUTPUT <br> VOLTAGE (V) | GAIN <br> $\left(\mathrm{V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ | GAIN IN dB <br> $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ |
| :--- | :---: | :---: | :---: | :---: |
|  |  |  |  |  |

## RESULT:

Bandwidth $\mathrm{BW}=\mathrm{f}_{2}-\mathrm{f}_{1}$
theoretical frequency $f=1 / 2 \pi \sqrt{ } \mathrm{C} L$

## CONCLUSION:

## VIVA QUESTIONS:

1. What is tuned amplifier? What are the various types of tuned amplifiers?
2. Define tuned amplifier?
3. Why tuned amplifier cannot be used at low frequency?
4. What is the other name for tuned amplifier?
5. Mention the Two Applications of tuned amplifiers?

## CIRCUIT DIAGRAM:

ASTABLE MULTIVIBRATOR Using (BJT):


ASTABLE MULTIVIBRATOR Using MOSFET:


## 2N7OOO MOSFET Pinout



## ASTABLE MULTIVIBRATOR

AIM: To Design an Astable Multivibrator to generate a square wave of 1 KHz frequency using Transistor.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | BJT \& MOSFET | BC 107 (or )2N7000 | 2 |
| 2 | Resistors | $1 \mathrm{~K} \Omega, 10 \mathrm{~K} \Omega(\mathrm{BJT})$ | 2 |
|  |  | $20 \mathrm{k} \Omega, 10 \mathrm{k} \Omega, 470 \Omega(\mathrm{MOSFET})$ | 1 |
| 3 | Capacitors | $1 \mu \mathrm{~F}$ | 2 |
|  |  | $0.1 \mu \mathrm{~F}$ | 2 |
| 4 | Function generator | $0-30 \mathrm{MHz}$ | 1 |
| 5 | RPS | $0-30 \mathrm{~V}$ | 1 |
| 6 | CRO | 30 MHz | 1 |
| 7 | Bread board | - | 1 |
| 8 | Connecting wires |  | As Per Required |

## OPERATION:

When the power is applied, due to some imbalance in the circuit, the transistor $\mathrm{Q}_{2}$ conducts more than $\mathrm{Q}_{1}$ i.e. current flowing through transistor $\mathrm{Q}_{2}$ is more than the current flowing in transistor $\mathrm{Q}_{1}$. The voltage $\mathrm{V}_{\mathrm{C} 2}$ drops. This drop is coupled by the capacitor $\mathrm{C}_{1}$ to the base by $\mathrm{Q}_{1}$ there by reducing its forward baseemitter voltage and causing $\mathrm{Q}_{1}$ to conduct less. As the current through $\mathrm{Q}_{1}$ decreases, $\mathrm{V}_{\mathrm{c} 1}$ rises. This rise is coupled by the capacitor $\mathrm{C}_{2}$ to the base of $\mathrm{Q}_{2}$. There by increasing its base- emitter forward bias. This $\mathrm{Q}_{2}$ conducts more and more and $\mathrm{Q}_{1}$ conducts less and less, each action reinforcing the other. Ultimately $\mathrm{Q}_{2}$ gets saturated and becomes fully ON and $\mathrm{Q}_{1}$ becomes OFF. During this time $\mathrm{C}_{1}$ has been charging towards $V_{c c}$ exponentially with a time constant $T_{1}=R_{1} C_{1}$. The polarity of $C_{1}$ should be such that it should supply voltage to the base of $\mathrm{Q}_{1}$. When $\mathrm{C}_{1}$ gains sufficient voltage, it drives $\mathrm{Q}_{1} \mathrm{ON}$. Then VC1 decreases and makes $\mathrm{Q}_{2}$ OFF. $\mathrm{V}_{\mathrm{C} 2}$ increases and makes $\mathrm{Q}_{1}$ fully saturated. During this time $\mathrm{C}_{2}$ has been charging through $\mathrm{V}_{\mathrm{CC}}, \mathrm{R}_{2}, \mathrm{C}_{2}$ and $\mathrm{Q}_{2}$ with a time constant $\mathrm{T} 2=\mathrm{R}_{2} \mathrm{C}_{2}$. The polarity of $\mathrm{C}_{2}$ should be such that it should supply voltage to the base of $\mathrm{Q}_{2}$. When $\mathrm{C}_{2}$ gains sufficient voltage, it drives $\mathrm{Q}_{2} \mathrm{On}$, and the process repeats.

## THEORETICAL CALCULATIONS:

$$
\begin{gathered}
\mathrm{F}=1 / \mathrm{T}=(1 / 1.38 \mathrm{RC}) \\
\mathrm{R}=10 \mathrm{~K} \Omega \quad \mathrm{C}=0.1 \mu_{\mathrm{F}}
\end{gathered}
$$

## MODEL GRAPH:




TABULAR COLUMN:

| Practical values | Amplitude (v) | Practical Time period | Practical $\mathrm{F}(\mathrm{hz})$ |
| :--- | :--- | :--- | :--- |
| $\mathrm{V}_{\mathrm{C} 1}=$ Voltage at collector of $\mathrm{Q}_{1}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{B} 2}=$ Voltage at base of $\mathrm{Q}_{2}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{C} 2}=$ Voltage at collector of $\mathrm{Q}_{2}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{B} 1}=$ Voltage at base of $\mathrm{Q}_{1}$ |  |  |  |

## Design Procedure:

The period T is given by
$\mathrm{T}=\mathrm{T}_{1}+\mathrm{T}_{2}=0.69\left(\mathrm{R}_{1} \mathrm{C}_{1}+\mathrm{R}_{2} \mathrm{C}_{2}\right)$
For symmetrical circuit, with $\mathrm{R}_{1}=\mathrm{R}_{2}=\mathrm{R} \& \mathrm{C}_{1}=\mathrm{C}_{2}=\mathrm{C}$
$\mathrm{T}=1.38 \mathrm{RC}$
Let $\mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V} ; \mathrm{h}_{\mathrm{fe}}=51$ (for BC 107 ), $\mathrm{V}_{\mathrm{BESat}}=0.7 \mathrm{~V} ; \mathrm{V}_{\mathrm{CESat}}=0.3 \mathrm{~V}$ Let $\mathrm{C}=0.1 \mu \mathrm{~F} \& \mathrm{~T}=1 \mathrm{mSec}$.
$10^{-3}=1.38 \times \mathrm{RX} \mathrm{0.1} \mathrm{X} 10^{-6}$
$\mathrm{R}=7.24 \mathrm{~K} \Omega$ (Practically choose $10 \mathrm{~K} \Omega$ ) i.e., R1 and R2 resistors.
Let $\mathrm{I}_{\text {Cmax }}=10 \mathrm{~mA}$
$\mathrm{R}_{\mathrm{c}}=\frac{V_{c c}-V_{\text {cesat }}}{I_{\text {cmax }}}=\quad=\frac{12-0.3}{0.01}$
$R_{c}==1.17 \mathrm{~K} \Omega(1 \mathrm{~K} \Omega$ is selected for Rc 1 and Rc 2$)$

## PROCEDURE:

1. Make then connections as per the circuit diagram.
2. Observe the Base Voltage and Collector Voltages of Q1 \& Q2 on CRO in DC mode and measure the frequency $(f=1 / T)$.
3. Trace the waveforms at collector and base as each transistor with the help of dual trace CRO and plot the waveforms.
4. Verify the practical output frequency with theoretical values $\mathrm{f}=1 / \mathrm{T}$, where $\mathrm{T}=1.38 \mathrm{RC}$

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. What are the other names of Astable multivibrator?
2. Define quasi stable state?
3. Is it possible to change time period of the waveform without changing $\mathrm{R} \& \mathrm{C}$ ?
4. Explain charging and discharging of capacitors in an Astable Multivibrator?
5. How can an Astable multivibrator be used as VCO?

CIRCUIT DIAGRAM: MONOSTABLE MULTIVIBRATOR


## MODEL WAVEFORMS:



## $\operatorname{Exp}$ No: 11

## Date:

## MONOSTABLE MULTIVIBRATOR

AIM: To Design a Monostable Multivibrator for the pulse width of 0.3 msec

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :--- | :---: | :---: |
| 1 | TRANSISTOR | BC 107 | 2 |
| 2 | Resistors | $100 \mathrm{~K} \Omega$ | 1 |
|  |  | $47 \mathrm{~K} \Omega$ | 1 |
|  |  | $10 \mathrm{~K} \Omega$ | 2 |
|  |  | $1 \mathrm{~K} \Omega$ | 2 |
| 3 | Capacitors | $0.1 \mu \mathrm{~F}$ | 2 |
|  |  | $0.01 \mu \mathrm{~F}$ | 1 |
| 5 | Diode | IN 4007 | 1 |
| 6 | Runction generator | $0-30 \mathrm{MHz}$ | 1 |
| 7 | CRO | $0-30 \mathrm{~V}$ | 1 |
| 8 | Bread board | 30 MHz | 1 |
| 9 | Connecting wires | - | 1 |

## OPERATION:

Assume initially transistor $\mathrm{Q}_{2}$ is in saturation as it gets base bias from $\mathrm{V}_{\mathrm{CC}}$ through R . coupling from $\mathrm{Q}_{2}$ collector to $\mathrm{Q}_{1}$ base ensures that Q 1 is in cutoff. If an appropriate negative trigger pulse applied at collector of $\mathrm{Q}_{1}\left(\mathrm{~V}_{\mathrm{C} 1}\right)$ induces a transition in $\mathrm{Q}_{2}$, then Q 2 goes to cutoff. The output at $\mathrm{Q}_{2}$ goes high. This high output when coupled to $\mathrm{Q}_{1}$ base, turns it ON . The $\mathrm{Q}_{1}$ collector voltage falls by $\mathrm{I}_{\mathrm{C} 1} \mathrm{R}_{\mathrm{C} 1}$ and Q 2 base voltage falls by the same amount, as voltage across a capacitor ' C ' cannot change instantaneously.

The moment, a negative trigger is applied at $\mathrm{V}_{\mathrm{C} 1}, \mathrm{Q}_{2}$ goes to cutoff and $\mathrm{Q}_{1}$ starts conducting. There is a path for capacitor $C$ to charge from $V_{C C}$ through $R$ and the conducting transistor $Q_{1}$. The polarity should be such that $\mathrm{Q}_{2}$ base potential rises. The moment, it exceeds $\mathrm{Q}_{2}$ base cut-in voltage, it turns ON $\mathrm{Q}_{2}$ which due to coupling through R1 from collector of Q 2 to base of Q 1, turns Q1 OFF. Now we are back to the original state i.e. $\mathrm{Q}_{2}$ is ON and Q1 is OFF. Whenever trigger the circuit into the other state, it cannot stay there permanently and it returns back after a time period decided by $R$ and C. Pulse width is given as $\mathrm{T}=0.69 \mathrm{RCsec} . \mathrm{F}$

TABULAR COLUMN:

| Practical values | Amplitude (v) | Practical Time period | Practical $\mathrm{F}(\mathrm{hz})$ |
| :--- | :--- | :--- | :--- |
| $\mathrm{V}_{\mathrm{C} 1}=$ Voltage at collector of $\mathrm{Q}_{1}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{B} 2}=$ Voltage at base of $\mathrm{Q}_{2}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{C} 2}=$ Voltage at collector of $\mathrm{Q}_{2}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{B} 1}=$ Voltage at base of $\mathrm{Q}_{1}$ |  |  |  |

## THEORETICAL CALCULATIONS:

$\mathrm{T}_{\mathrm{ON}}=0.69 \mathrm{RC} \quad \mathrm{R}=47 \mathrm{~K} \% \mathrm{u} 2126$ and $\mathrm{C}=10 \mathrm{nF}$ or $0.01 \mu \mathrm{~F}$

## Design Procedure:

To design a monostable multivibrator for the Pulse width of 0.3 mSec .
Let $\mathrm{I}_{\mathrm{Cmax}}=15 \mathrm{~mA}, \mathrm{~V}_{\mathrm{CC}}=15 \mathrm{~V}, \mathrm{~V}_{\mathrm{BB}}=15 \mathrm{~V}, \mathrm{R}_{1}=10 \mathrm{~K} \% \mathrm{u} 2126 . \quad \mathrm{T}=0.69 \mathrm{RC}$
Choose $\mathrm{C}=10 \mathrm{nf}(0.01 \mu \mathrm{~F}) \quad \mathrm{T}=0.69 \mathrm{RC}$

$$
0.3 \times 10^{-3} \mathrm{Sec}=0.69 \times \mathrm{R} \times 10 \times 10^{-9}
$$

$\mathrm{R}=43.47 \mathrm{Kohms} \approx 47 \mathrm{Kohms}$

$$
\mathrm{R}_{\mathrm{C}}=\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{CESAT}) / \mathrm{I}_{\mathrm{CMAX}}}=(15-0.3) / 15 \times 10^{-3}=1 \mathrm{~K} \mathrm{ohms}\right.
$$

Minimum requirement of $\left|V_{\mathrm{B} 1}\right| \leq 0.1$
For more margin, given $V_{B 1}=-1.185$

$$
\mathbf{V}_{\mathrm{B} 1}=\frac{-V_{B B}}{R 1+R 2} \mathbf{R} 1+\frac{-V_{\text {CEsat }}}{R 1+R 2} \mathbf{R} 2
$$

Substitute the values, $\mathrm{R}_{1}=10 \mathrm{kohms}$ we will get $\mathrm{R} 2=100 \mathrm{Kohms}$

## PROCEDURE:

1. Make the connections as per the circuit diagram.
2. Select the triggering pulse such that the frequency is less than $1 / T$
3. Apply the triggering input to the circuit and to the CRO's channel and Connect the CRO channel-2 to the collector and base of the Transistor Q1\&Q2.
4. Adjust the triggering pulse frequency to get stable pulse on the CRO and now measure the pulse width and verify with the theoretical value.
5. Obtain waveforms at different points like $\mathrm{V}_{\mathrm{B} 1}, \mathrm{~V}_{\mathrm{B} 2}, \mathrm{~V}_{\mathrm{C} 1} \& \mathrm{~V}_{\mathrm{C} 2}$ and plot the graph.

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. What is a multivibrator?
2. What are applications of Monostable Multivibrator?
3. The monostable multivibrator is also called as -----------
4. A Monostable Multivibrator generates wave
5. Why is the time period T also called Delay time?

CIRCUIT DIAGRAM: Bistable MultiVibrator:


## MODEL WAVEFORM:



## Date:

## BITABLE MULTIVIBRATOR

AIM: To Design Bistable Multivibrator to generate a square wave of 55 KHz frequency using Transistor.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | TRANSISTOR | $(\mathrm{BC} \mathrm{107)}$ | 2 |
| 2 |  | $1 \mathrm{~K} \Omega$ | 4 |
|  |  | $100 \mathrm{~K} \Omega$ | 2 |
| 3 |  | $0.001 \mu \mathrm{~F}, 0.33 \mu \mathrm{~F}$ | 2 |
| 4 | Diode | IN 4007 | 4 |
| 5 | Function generator | $0-30 \mathrm{MHz}$ | 1 |
| 6 | RPS | $0-30 \mathrm{~V}$ | 1 |
| 7 | CRO | 30 MHz | 1 |
| 8 | Bread board | - | 1 |
| 9 | Connecting wires | - | As Per Required |

## OPERATION:

When VCC is applied, one transistor will start conducting slightly more than that of the other, because of some differences in the characteristics of a transistor. Let $\mathrm{Q}_{2}$ be ON and $\mathrm{Q}_{1}$ be OFF. When $\mathrm{Q}_{2}$ is ON , The potential at the collector of $\mathrm{Q}_{2}$ decreases, which in turn will decrease the potential at the base of $\mathrm{Q}_{1}$ due to potential divider action of $\mathrm{R}_{1}$ and $\mathrm{R}_{2}$. The potential at the collector of $\mathrm{Q}_{1}$ increases which in turn further increases the base to emitter voltage at the base of $\mathrm{Q}_{2}$. The voltage at the collector of $\mathrm{Q}_{2}$ further decreases, which in turn further reduces the voltage at the base of $Q_{1}$. This action will continue till $\mathrm{Q}_{2}$ becomes fully saturated and $\mathrm{Q}_{1}$ becomes fully cutoff.

Thus the stable state of binary is such that one device remains in cut-off and other device remains at saturation. It will be in that state until the triggering pulse is applied to it. It has two stable states. For every transition of states triggering is required. At a time only one device will be conducting

## THEORETICAL CALCULATIONS:

## PROCEDURE:

1. Make the connections as per the circuit diagram.
2. Apply trigger pulse of $1 \mathrm{KHz} 5 \mathrm{v}(\mathrm{p}-\mathrm{p})$ from function generator.
3. Obtain waveforms at different points such as $\mathrm{V}_{\mathrm{B} 1}, \mathrm{~V}_{\mathrm{B} 2}, \mathrm{~V}_{\mathrm{C} 1} \& \mathrm{~V}_{\mathrm{C} 2}$.
4. Trace the waveform at collector and base of each transistor with the help of dual trace CRO. Note the Time relation of waveform

## Design Procedure:

$$
\begin{aligned}
\mathbf{R}_{\mathbf{C}} & =\frac{\left(V_{C C}-V_{\text {CEsat }}\right)}{I_{\text {Cmax }}} \\
& =\frac{15-0.3}{15 * 10^{-3}}=1 \mathrm{k} \Omega
\end{aligned}
$$

$\mathbf{V}_{\mathrm{B} 1}=\frac{-V_{B B}}{R 1+R 2} \mathbf{R} 1+\frac{-V_{\text {CEsat }}}{R 1+R 2} \mathbf{R} 2$
$-1.2=\left(-15 \mathrm{R}_{1}+0.2 \mathrm{R}_{2}\right) /\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right) ;$ given $\mathrm{R} 1=10 \mathrm{~K} \Omega$
$\mathrm{R} 2=100 \mathrm{~K} \Omega$
$\mathrm{R}_{1}=10 \mathrm{~K} \Omega, \mathrm{R}_{2}=100 \mathrm{~K} \Omega$ and $\mathrm{C}=0.1 \mu \mathrm{~F}$
$\mathrm{F}_{\text {max }}=\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right) / 2 \mathrm{C} \mathrm{R}_{1} \mathrm{R}_{2}$

$$
=(10+100) \times 10^{3} /\left(2 \times 0.3 \times 10^{-6} \times 10 \times 100 \times 10^{6}\right)=55 \mathrm{KH}
$$

TABULAR COLUMN:

| Practical values | Amplitude (v) | Practical Time period | Practical F(hz) |
| :--- | :--- | :--- | :--- |
| $\mathrm{V}_{\mathrm{C} 1}=$ Voltage at collector of $\mathrm{Q}_{1}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{B} 2}=$ Voltage at base of $\mathrm{Q}_{2}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{C} 2}=$ Voltage at collector of $\mathrm{Q}_{2}$ |  |  |  |
| $\mathrm{~V}_{\mathrm{B} 1}=$ Voltage at base of $\mathrm{Q}_{1}$ |  |  |  |

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. What are the other names of Bistable Multivibrator?
2. What are the applications of a Bistable Multivibrator?
3. Describe the operation of commutating capacitors?
4. Commutating capacitors are also called as $\qquad$ or $\qquad$ .
5. What is the meaning of a stable state in a multi-vibrator?

## ADDITIONAL EXPERIMENTS

## CIRCUIT DIAGRAM:

## CLASS-B PUSHPULL POWER AMPLIFIER



MODFEL WAVEFORM:


## Expt No: 01

## Date:

## CLASS-B PUSHPULL POWER AMPLIFIER

AIM: To design a class-B push-pull power amplifier in order to achieve maximum output AC power and efficiency.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :---: | :---: | :---: |
| 1 | Transistor | $(\mathrm{BC} 107)$ | 2 |
| 2 |  | $15 \Omega$ | 1 |
|  |  | $100 \Omega$ | 1 |
|  | Resistors | $6.2 \mathrm{~K} \Omega$ | 1 |
|  |  | $100 \Omega($ Speaker |  |
| 3 | Zener Diode | IN 4730 A | 1 |
| 4 | RPS | $0-30 \mathrm{~V}$ | 1 |
| 5 | Function Generator | $0-3 \mathrm{MHz}$ | 1 |
| 6 | CRO | $0-30 \mathrm{MHz}$ | 1 |
| 7 | Bread board | - | 1 |
| 8 | Connecting wires | - | As Per Required |

## OPERATION:

The circuit arrangement of the Class B push pull amplifier is similar to the Class A push pull amplifier except for the absence of the biasing resistors. T1 is the input coupling capacitor and the input signal is applied to its primary. Q1 and Q2 are two identical transistors and their emitter terminals are connected together. Center tap of the input coupling transformer and the negative end of the voltage source is connected to the junction point of the emitter terminals. Positive end of the voltage source is connected to the center tap of the output coupling transformer. Collector terminals of each transistor are connected to the respective ends of the primary of the output coupling transformer T2. Load RL is connected across the secondary of T2.

The input signal is converted into two similar but phase opposite signals by the input transformer T1. One out of these two signals is applied to the base of the upper transistor while the other one is applied to the base of the other transistor. You can understand this from the circuit diagram.

When transistor Q1 is driven to the positive side using the positive half of its input signal, the reverse happens in the transistor Q2. That means when the collector current of Q1 is going in the increasing direction, the collector current of Q2 goes in the decreasing direction. Anyway the current flow through the respective halves of the primary of the T 2 will be in same direction. Have a look at the figure for better understanding. This current flow through the T 2 primary results in a wave form induced across its secondary. The wave form induced across the secondary is similar to the original input signal but amplified in terms of magnitude.

## PROCEDURE:

1. Connect the circuit diagram as shown in the figure.
2. Determine the maximum signal handling capacity of the push pull amplifier.
3. Apply sinusoidal signal of 1 V peak to peak voltage at a frequency of 1 kHz .
4. Connect the loud speaker at the output. If you hear sound note the circuit is working .Now replace the loud speaker with the Power meter at the $\mathrm{O} / \mathrm{P}$. Select the typical value of the load impedance in the power meter. Also connect the output to a CRO. See the wave shape of the output voltage on the CRO. Increase the input signal voltage till the output wave shape starts getting distorted. Note this input signal voltage . now reduce the input signal to a value slightly below this voltage.
5. By changing the load impedance note down the value of impedance and output Power meter.
6. Tabulate the readings.
7. Plot the graph between output Power and load impedance.
8. Select the load impedance is equal to the optimum load. See the Wave shape of the output on the CRO. Increase the input signal till the wave shape shows distortion and note down the input voltage. This gives the maximum signal handling capacity of the amplifier.

## OBSERVATION:

$\mathrm{VCC}=12 \mathrm{~V}$
$R L=100 \Omega$
EFFICIENCY: Pac/Pdc $=\mathrm{V}_{\mathrm{mx}} \mathrm{II} / 4 \mathrm{X} \mathrm{Vcc}=$
$\mathrm{Vm}=\mathrm{Vpp} / 2$
$\mathrm{V}_{\mathrm{S}}=20 \mathrm{mV}$
Input power, $\mathrm{Pin}=2 \mathrm{VccIm} /$ п
Output power, Pout=VmIm/2
Power Gain or efficiency, $\eta=\Omega / 4(\mathrm{Vm} / \mathrm{Vcc}) 100$

## OBSERVATIONS:

$\mathrm{V}_{\mathrm{S}}=20 \mathrm{mV}$

| S.NO | RL | Output Power in db <br> $\mathrm{P}_{\mathrm{O}}(\mathrm{mW})\left(10 \log \mathrm{P}_{0}\right)$ | Power in db $(10 \log \mathrm{P} 0)$ |
| :---: | :---: | :---: | :---: |
|  |  |  |  |

## RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. What is the conduction angle of push pull amplifier?
2. What is the power efficiency of push pull amplifier?
3. What is crossover distortion?
4. What is operating point in push pull amplifier?
5. What are the advantages \& disadvantages of class B Push Pull Amplifier?

## CIRCUIT DIAGRAM:

## CASCADE AMPLIFIER:



MODEL WAVEFORM:


Expt No: 02
Date:
CASCADE AMPLIFIER

AIM: To obtain the Voltage gain for Cascade Amplifier and also to observe the frequency Response.

## APPARATUS:

| S.NO | APPARATUS | RANGE | QUANTITY |
| :---: | :--- | :--- | :---: |
| 1 | Transistor | $(\mathrm{BC} 107$ or BC547 ) | 2 |
| 2 | 2 | Resistors | $10 \mathrm{~K} \Omega$ |
|  |  | 2 |  |
|  |  | $1 \mathrm{~K} \Omega$ | 2 |
|  |  | $3.3 \mathrm{~K} \Omega$ | 2 |
|  | $220 \Omega$ | 2 |  |
| 3 | Capacitors | $10 \mu \mathrm{~F}$ | 5 |
| 4 | RPS | $0-12 \mathrm{~V}$ | 1 |
| 5 | Function Generator | $0-3 \mathrm{MHz}$ | 1 |
| 6 | CRO | 30 MHz | 1 |
| 7 | Bread board | - | 1 |
| 8 | Connecting wires | - | As Per Required |

## PROCEDURE:

1) Connect the circuit as shown in the figure.
2) Apply 1 Khz frequency and 20 mv Vp-p Sine wave from function generator..
3) Observe input and output Waveforms simultaneously on C.R.O
4) Change the frequency of input signal from 10 HZ to 1 MHZ in steps and note amplitudes of input and output Waveforms(input signal should be maintained constant).
5) Calculate Voltage gain (A) for each (in db) verses frequency.

## PRECAUTIONS:

Avoid loose connections give proper input voltage

## MODEL GRAPH:

Input wave form


## Output waveform




| S.NO | FREQUENCY(Hz) | OUTPUT <br> VOLTAGE (V) | GAIN <br> $\left(\mathrm{V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ | GAIN IN dB <br> $\mathrm{A}_{\mathrm{v}}=20 \log _{10}\left(\mathrm{~V}_{0} / \mathrm{V}_{\mathrm{i}}\right)$ |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |
|  |  |  |  |  |

## CALCULATION:

Maximum gain of the amp:
Upper cutoff frequency F2:
Lower cutoff frequency F1:
Band width=F2-F1

RESULT:

## CONCLUSION:

## VIVA QUESTIONS:

1. What is a cascaded amplifier?
2. What is the voltage gain of cascade amplifier?
3. What is Current gain of cascade amplifier?
4. What is the input and output impedance of Cascade Amplifier?
5. What is the need of Cascade Amplifier?
